

Figure 3. Loss vs. frequency for 50  $\Omega$  microstrip transmission lines on three substrates: 250  $\mu\text{m}$  Duroid, 100  $\mu\text{m}$  GaAs, 13  $\mu\text{m}$  polyimide. The polyimide loss is measured data, the others are calculations. Peaks in the measured data result from the measured line being an integer-multiple of a half-wavelength.

Figure 3 shows a comparison between the losses of 50  $\Omega$  microstrip transmission lines using three different dielectrics with their typical thicknesses: 250  $\mu\text{m}$  Duroid [6], 100  $\mu\text{m}$  GaAs, 13  $\mu\text{m}$  polyimide [7]. For good dielectrics, the loss of a microstrip line is dominated by metal resistivity, and since a wider metal line is needed for thicker substrates to maintain the desired impedance, a thicker substrate will generally provide lower loss [5].

### Polyimide Microstrip MMICs

Polyimide microstrip transmission lines consist of a metal ground-plane that is the top-most layer of metal on a standard processed Si IC wafer, and a thick layer of polyimide which supports the microstrip line (figure 1) [5]. This structure is then a conventional MMIC using the components from a standard Si IC process (transistors, resistors, capacitors, diodes, etc.) under the ground plane with the transmission line elements above. Such a structure makes highly integrated circuits possible, for example, by including microwave circuits (low-noise amplifiers, mixers, oscillators, etc.) on the same chip as signal processing and digital circuitry.

We are currently working with IBM using their 200 mm SiGe bipolar technology [2]. In this variation of a standard Si IC process, we have three levels of metalization normally used

for interconnections. We use the top-most layer of metal as a ground-plane for the microstrip transmission lines, but the ground-plane must have openings to allow connections between the lines above the and the devices below.

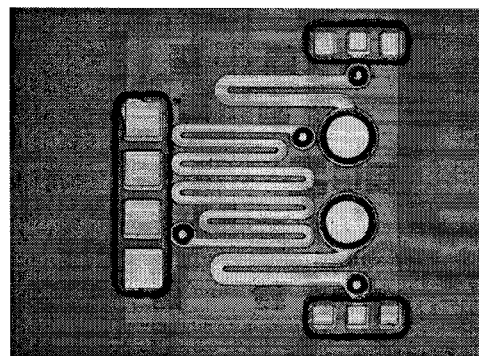


Figure 4. A Ku-Band driver amplifier using polyimide microstrip transmission lines. Large circles are capacitors and small circles are vias between devices under the ground-plane and the microstrip lines.

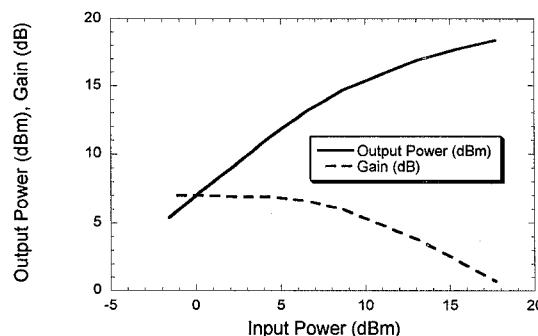


Figure 5. Output power and gain vs. input power for amplifier shown in figure 4. Small-signal gain is 7 dB, and 1 dB compression occurs at 40 mW output power (12 GHz).

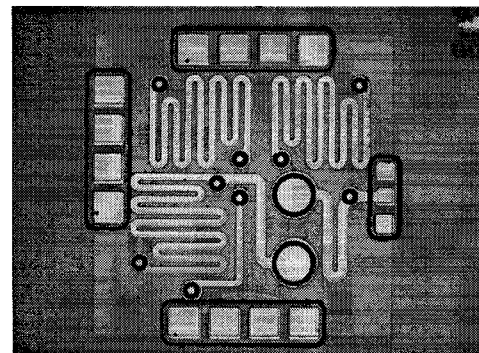


Figure 6. Photograph of a Ku-Band VCO with buffer amplifier. Distributed elements are implemented using polyimide microstrip transmission lines.

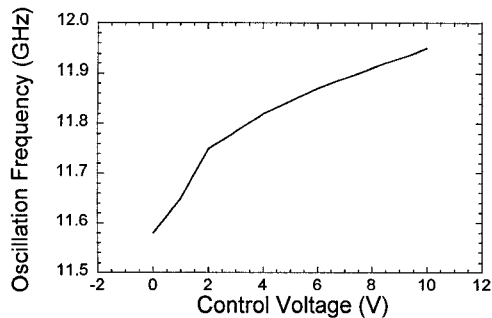


Figure 7. Frequency vs. control voltage for the VCO shown in figure 6.

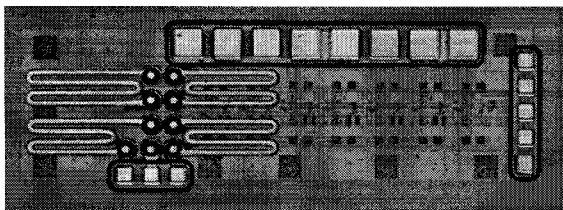


Figure 8. A K-band static frequency divider ( $\div 128$ ) implemented with an inductively peaked input buffer (left side of photo).

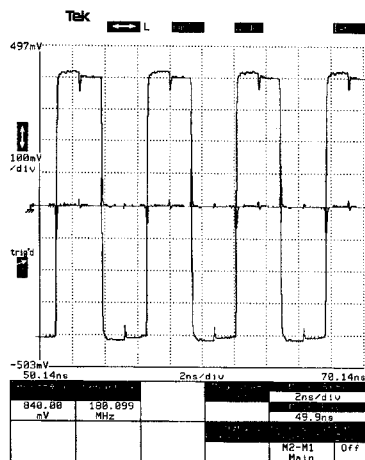


Figure 9. Response of the frequency divider ( $\div 128$ ) shown in figure 8. The maximum operating frequency is 23 GHz.

The substrate, devices, interconnects, and the ground-plane are all part of the standard IC process.

We have designed and tested a variety of polyimide MMICs that combine the IBM SiGe HBT devices [2,3] with polyimide dielectric microstrip lines [5]. Figures 4 through 9 show photographs and measured responses of three

SiGe MMICs: a Ku-Band amplifier, a Ku-Band VCO, and a K-band frequency divider [8]. An X-band singly-balanced mixer was also reported at [9].

### Flip-Chip MICs

Flip-chip IC attachment has been widely used in the IC industry recently to reduce the parasitics associated with the more typical wire bonding to packages. Flip-chip attachment requires one to place a solder bump on the IC's pads. These solder bumps are then placed in contact with their desired pads on the package and heated to melt the solder, forming the connection (figure 2).

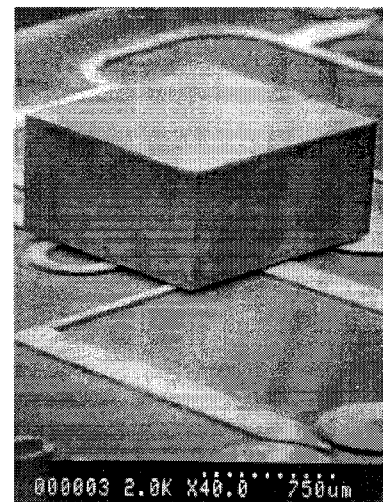


Figure 10. SEM photograph of a Si die flip-chip mounted to a Duroid circuit board. The circuit board contains transmission lines and other distributed elements. This chip is  $1300 \times 1300 \times 750 \mu\text{m}^3$ .

We use the same flip-chip technique to attach small die (typically discrete transistors) to microwave circuit boards. Figure 10 shows a photograph of a small die flip-chip mounted on a Duroid circuit board. Wire bonding has a large parasitic inductance that can change from bond to bond even in automated systems. The flip-chip MIC combines the repeatable, low parasitic attachment of solder bumps with the small die size and high performance one achieves with the MIC.

The flip-chip MIC approach also allows one to incorporate a variety of technologies in a single circuit. For example, one could use an InP-based HEMT for an LNA, and on the same circuit have a high-speed SiGe A-D converter.

Many more combinations are possible that can take advantage of the combination of technologies.

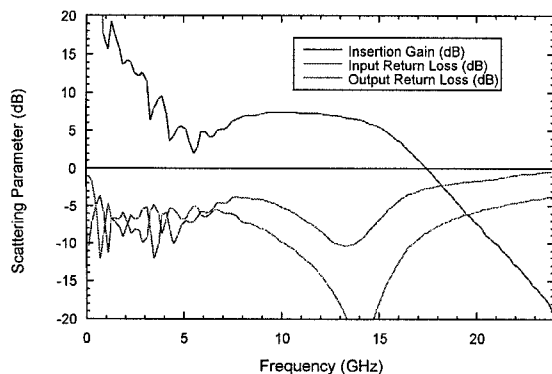


Figure 11. Small-signal response of Ku-Band flip-chip MIC amplifier. The gain is 7 dB over the band of interest.

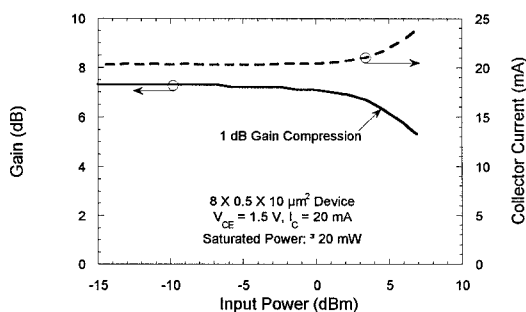


Figure 12. Large-signal response of Ku-Band flip-chip MIC amplifier. 1 dB gain compression occurs at 20 mW output power,  $\approx \frac{1}{2}$  of that from the amplifier in figure 4 (which uses a device 2.5 X larger).

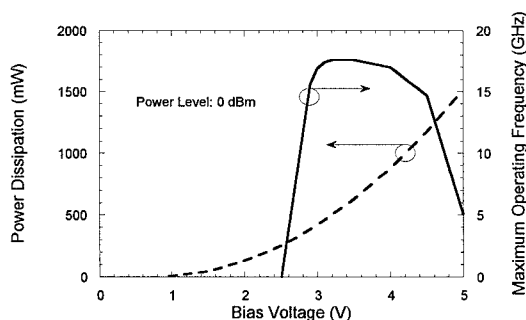


Figure 13. Response of the Ku-Band frequency divider flipped onto a Duroid substrate. This circuit is different from that shown in figure 8 only in that it does not have inductive peaking at the input buffer and uses two additional divider stages ( $\div 512$ ).

We have designed and tested several SiGe flip-chip MIC circuits combining the IBM SiGe HBT with passive elements fabricated on 10 mil RO3010 Duroid. Small- and large-signal performance of a Ku-Band amplifier are shown in figures 11 and 12. This amplifier is similar to that shown in figure 4, but with a much smaller device ( $40 \mu\text{m}^2$  compared to  $100 \mu\text{m}^2$ ). Characteristics of a Ku-Band frequency divider, similar to that shown figure 8, are shown in figure 13.

## Summary

We have demonstrated two approaches (polyimide MMIC and flip-chip MIC) for implementing low-loss distributed elements for integration with Si-based devices. There is a tradeoff that differentiates the two approaches. Although the loss of a transmission line on a thick Duroid board is much lower than that on thin polyimide, the parasitics associated with the lines on the Si-based die feeding the bumps and the bumps themselves impose difficult-to-model and often significant parasitics; while the parasitics associated with the vias in the polyimide MMIC are much smaller.

By combining the performance achievable with the MMIC or MIC approach and the low-cost of Si fabrication (as compared to III-V), the possibility of highly functional consumer microwave products may become reality. There are many applications that require high-volume, low-cost microwave components.

## References

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